



### On Moment Problems in Robust Control, Spectral Estimation, Image Processing and System Identification

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# Special recognition



Christopher Byrnes



Tryphon Georgiou

+ Sergei Gusev, Alexei Matveev, Alexandre Megretski, Giorgio Picci, Per Enqvist, Johan Karlsson, Ryozo Nagamune, Anders Blomqvist, Vanna Fanizza, Enrico Avventi, Axel Ringh

### What is the talk about

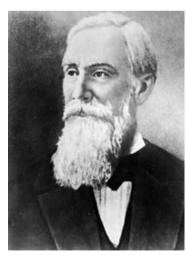
- A classical problem the moment problem with a decidedly non-classical twist motivated by engineering applications.
- What is new are certain rationality constraints imposed by applications that alter the mathematical problem and make it nonlinear.
- A global-analysis approach that studies the class of solutions as a whole.
- A powerful paradigm for smoothly parameterizing, comparing, and shaping solutions to specifications.

## The moment problem

Given 
$$c_0, c_1, \ldots, c_n$$
, find  $d\mu$  such that 
$$\int_a^b \alpha_k(t) d\mu = c_k, \quad k = 0, 1, \ldots, n$$
 space of positive

 $d\mu \in \mathcal{M}_+$ positive measures

- Power moment problem:  $\alpha_k(t) = t^k$
- Trigonometric moment problem:  $\alpha_k(t)=e^{ikt}, \quad [a,b]=[-\pi,\pi]$  Nevanlinna-Pick interpolation:  $\alpha_k(t)=\frac{e^{it}+z_k}{e^{it}-z_k}, \quad [a,b]=[-\pi,\pi]$



Chebyshev



Markov



Lyapunov

$$\int_a^b \alpha_k(t) \frac{d\mu}{d\mu} = \frac{c_k}{c_k}, \quad k = 0, 1, \dots, n$$

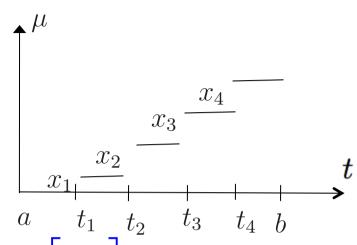
Ex. 1  $\mu$  step function  $\longrightarrow$ 

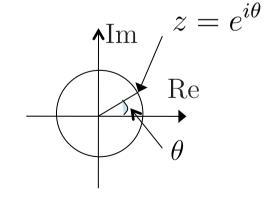
$$\sum_{j=1}^{N} \alpha_k(t_j) \mathbf{x_j} = \mathbf{c_k}, \quad k = 0, 1, \dots, n$$

$$\begin{bmatrix} \alpha_0(t_1) & \alpha_0(t_2) & \cdots & \alpha_0(t_N) \\ \alpha_1(t_1) & \alpha_1(t_2) & \cdots & \alpha_1(t_N) \\ \vdots & \vdots & \ddots & \vdots \\ \alpha_n(t_1) & \alpha_n(t_2) & \cdots & \alpha_n(t_N) \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_N \end{bmatrix} = \begin{bmatrix} c_0 \\ c_1 \\ \vdots \\ c_n \end{bmatrix}$$
infinitely many solutions

Ex. 2  $d\mu = \Phi(e^{i\theta}) \frac{d\theta}{2\pi}$  spectral density

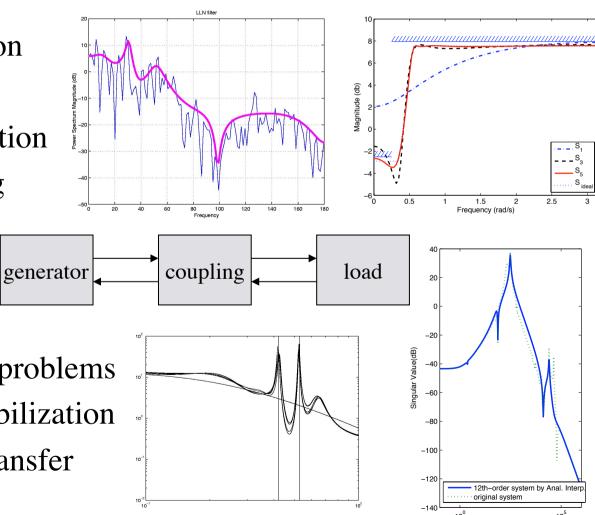
$$\int_{-\pi}^{\pi} e^{ik\theta} \Phi(e^{i\theta}) \frac{d\theta}{2\pi} = c_k, \quad k = 0, 1, \dots, n$$





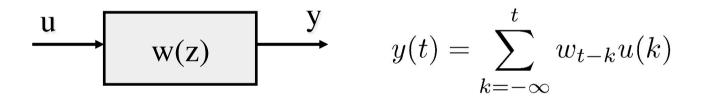
# Where do we find moment problems in applications?

- spectral estimation
- speech synthesis
- system identification
- image processing
- optimal control
- robust control
- model reduction
- model matching problems
- simultaneous stabilization
- optimal power transfer



# ... and why do these problems require a nonclassical approach?

• Solution must be of bounded complexity (such as rational of a bounded degree) so that one can realize it by a finite-dimensional device



System is finite-dimensional iff  $w(z) := \sum_{k=0}^{\infty} w_k z^{-k}$  is rational

• Classical theory does not provide natural parameterizations of rational solutions of bounded degree

### Prototype problem: Covariance extension

$$c_k = E\{y(t+k)y(t)\}, k = 0, 1, 2, ...,$$
  
where y stationary stochastic process

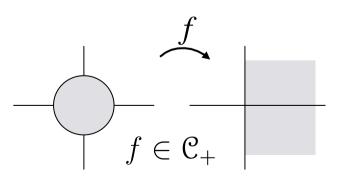
Given  $c_0, c_1, \ldots, c_n$ , find an infinite extension  $c_{n+1}, c_{n+2}, \ldots$  such that

Carathéodory Schur

$$f(z) = \frac{1}{2}c_0 + c_1z + \dots + c_nz^n + c_{n+1}z^{n+1} + \dots$$

(i) is a Carathéodory function

(ii) is rational of degree at most n



# Trigonometric moment problem

$$f(z) = \frac{1}{2}c_0 + c_1z + \dots + c_nz^n + c_{n+1}z^{n+1} + \dots$$

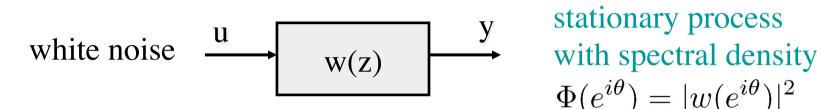
$$\Phi(e^{i\theta}) = 2\operatorname{Re}\{f(e^{i\theta})\} \\
= \sum_{k=-\infty}^{\infty} c_k e^{ik\theta} \ge 0 \\
c_{-k} = \bar{c}_k$$

### MOMENT PROBLEM: Find $\Phi$ of degree at most 2n such that

$$\frac{1}{2\pi} \int_{-\pi}^{\pi} e^{ik\theta} \Phi(e^{i\theta}) d\theta = c_k, \quad k = 0, 1, \dots, n$$

$$\Phi(z) = \frac{P(z)}{Q(z)}$$
,  $P, Q$  trigonometric polynomials of degree  $n$ 

# Spectral estimation by covariance extension



$$c_k = \frac{1}{2\pi} \int_{-\pi}^{\pi} e^{ik\theta} \Phi(e^{i\theta}) d\theta = E\{y(t+k)y(t)\}$$

$$= \lim_{N \to \infty} \frac{1}{N+1} \sum_{t=0}^{N-k} y_{t+k} y_t \qquad \text{where}$$

 $y_0, y_1, y_2, \ldots, y_N$ 

observed data

Since  $N < \infty$ , we use ergodic estimate

$$c_k = \frac{1}{N+1} \sum_{t=0}^{N-k} y_{t+k} y_t$$

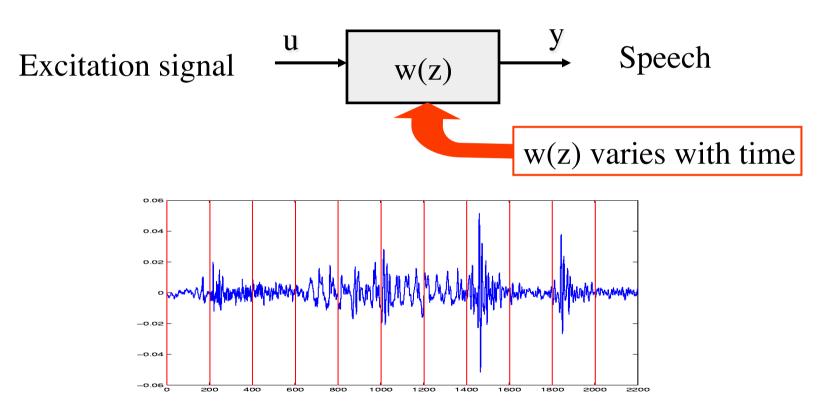
Hence, we can only estimate

$$c_0, c_1, \ldots, c_n, \quad n << N$$

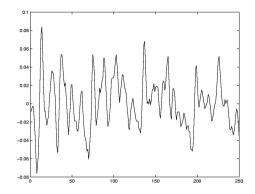
Remains to determine

$$c_{n+1}, c_{n+2}, c_{n+3}, \dots$$

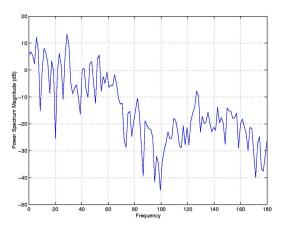
# Modeling speech



w(z) constant on each (30 ms) subinterval

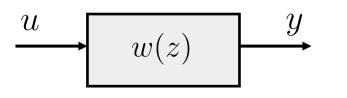


A 30 ms frame of speech for the voiced nasal phoneme [ng]

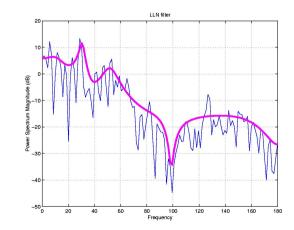


Periodogram (FFT) of voiced nasal phoneme [ng]

Construct a rational filter with a rational w(z) of low degree, modeling the window of speech



filter determined by few parameters



# Linear Predictive (LPC) Filtering

$$\begin{bmatrix} c_0 & c_1 & \cdots & c_{n-1} \\ c_1 & c_0 & \cdots & c_{n-2} \\ \vdots & \vdots & \ddots & \vdots \\ c_{n-1} & c_{n-2} & \cdots & c_0 \end{bmatrix} \begin{bmatrix} \varphi_{nn} \\ \varphi_{n,n-1} \\ \vdots \\ \varphi_{n1} \end{bmatrix} = \begin{bmatrix} c_n \\ c_{n-1} \\ \vdots \\ c_1 \end{bmatrix}$$

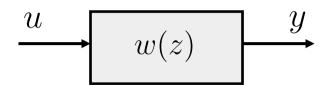
yields Szegö polynomial

$$\varphi_n(z) = z^n + \varphi_{n1}z^{n-1} + \dots + \varphi_{nn}$$

and modeling filter

$$w(z) = \frac{\sqrt{\rho_n}z^n}{\varphi_n(z)}$$
 where  $\rho_n = \sum_{j=0}^n c_j \varphi_{nj}$ 

# Linear Predictive (LPC) Filtering



$$w(z) = \sqrt{\rho_n} \frac{z^n}{\varphi_n(z)}$$

$$n = 10$$

Choose an excitation signal u from a code book with  $1024 = 2^{10}$  entries (10 bits)



Send coefficients in  $\varphi_{10}(z)$  and number of "best" signal



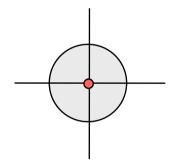
However, can we construct the general solution?

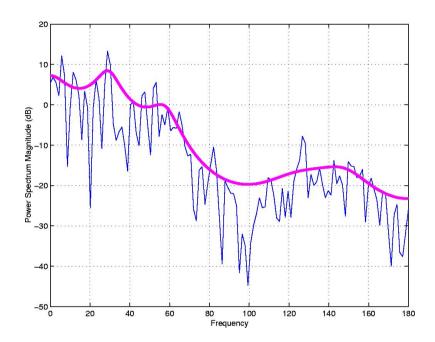
$$w(z) = \frac{\sigma(z)}{a(z)}$$

$$w(z) = \frac{\sigma(z)}{a(z)}$$

### Cellular telephone:

$$w(z) = \sqrt{\rho_n} \frac{z^n}{\varphi_n(z)}$$





FFT in blue envelope in purple

Is there a solution  $d\mu$  for each choice of spectral zeros?

YES (Georgiou 1983)

 $\sigma(z) = \sqrt{\rho_n} z^n$ 

spectral zeros

Unique? (Georgiou's conjecture)
Well-posed?

YES (Byrnes, Lindquist Gusev, Matveev 1993)

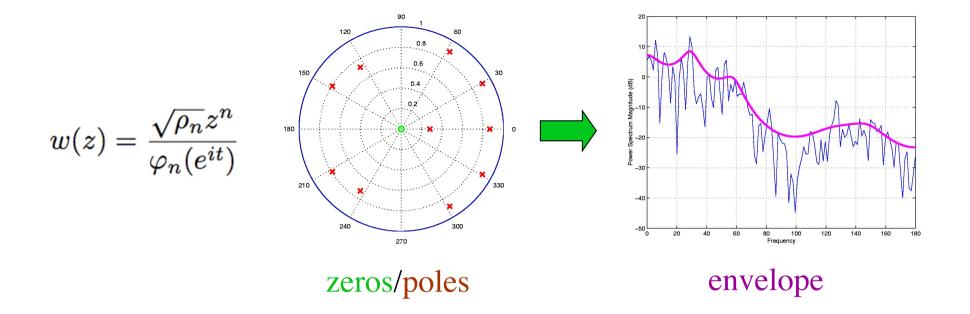
### All the solutions

THEOREM. The solutions of the rational covariance extension problem are completely parameterized by the zeros of the corresponding shaping filter, i.e., given an arbitrary monic stable polynomial  $\sigma(z)$  there is one and only one stable polynomial a(z) such that

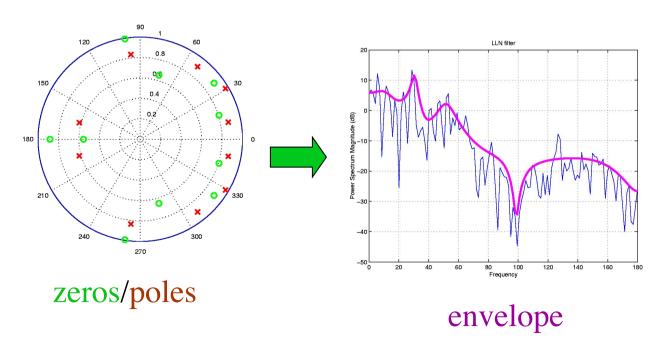
$$w(z) = \frac{\sigma(z)}{a(z)}$$

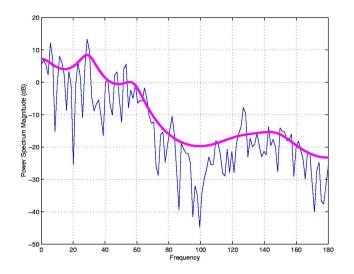
is a shaping filter for  $c_0, c_1, \ldots, c_n$ . The correspondence is a diffeomorphism.

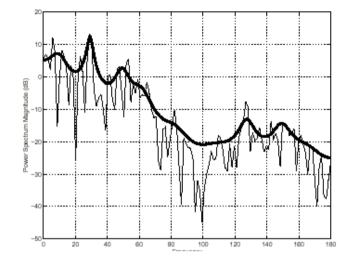
- Existence proved by Georgiou 1983; conjectured uniqueness
- The rest proved by Byrnes, Lindquist, Gusev, Matveev 1993
- These first proofs were nonconstructive, but there are now constructive proofs based on optimization



A w(z) with other spectral zeros, but with the same degree



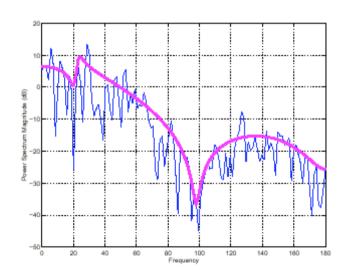


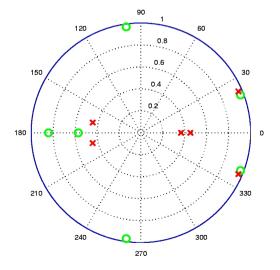


10th degree LPC

20th degree LPC

6th degree filter with appropriate zeros





# Optimization approach

Given  $\sigma(z)$  and c, minimize

$$\mathbb{J}_{\sigma}(q) = c_0 q_0 + c_1 q_1 + \dots + c_n q_n$$
$$-\frac{1}{2\pi} \int_{-\pi}^{\pi} |\sigma(e^{i\theta})|^2 \log Q(e^{i\theta}) d\theta$$

over all q such that

$$Q(e^{i\theta}) := q_0 + q_1 \cos \theta + \dots + q_n \cos n\theta > 0$$

THEOREM. There is a unique minimum.

$$f(z) = \frac{b(z)}{a(z)}$$

$$w(z) = \frac{\sigma(z)}{a(z)}$$

where 
$$|a(e^{i\theta})|^2 = Q(e^{i\theta})$$

Then 
$$f(z) = \frac{b(z)}{a(z)}$$
 where 
$$|a(e^{i\theta})|^2 = Q(e^{i\theta})$$
 
$$w(z) = \frac{\sigma(z)}{a(z)}$$
 
$$a(z)b(z^{-1}) + a(z^{-1})b(z) = \sigma(z)\sigma(z^{-1})$$

### Nonlinear coordinates

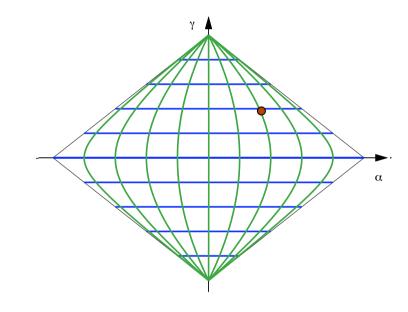
Normalize so that  $c_0 = 1$ 

The space of all rational Carathéodory functions f of degree at most n is a 2n-dimensional manifold.

A foliation with one leaf for each choice of  $\sigma$  (fast Kalman filtering)

A foliation with one leaf for each choice of  $c = (c_1, c_2, \dots, c_n)$ 

**THEOREM.** The two foliations intersect transversely so that each leaf in one meets each leaf in the other in exactly one point.

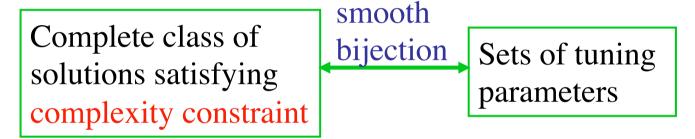


$$\min_{q} \mathbb{J}_{\sigma}(q)$$

unique solution  $\Phi = \frac{|\sigma|^2}{Q}$ 

# A global analysis approach

Find complete parameterization



- For any choice of tuning parameters, determine the corresponding solution by convex optimization
- Choose a solution that best satisfies additional design specifications (without increasing the complexity)

# Other applications leading to moment problem with rationality constraints

Given 
$$c_0, c_1, \ldots, c_n$$
, find  $d\mu$  such that 
$$\int_a^b \alpha_k(t) d\mu = c_k, \quad k = 0, 1, \ldots, n$$

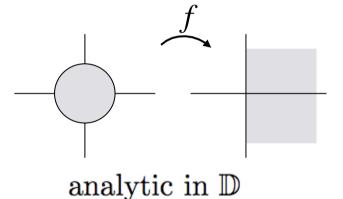
# Nevanlinna-Pick interpolation

$$\alpha_k(t) = \frac{e^{it} + z_k}{e^{it} - z_k}, \quad k = 0, 1, \dots, n \qquad z_0, z_1, \dots, z_n \in \mathbb{D}$$
 (distinct)

$$z_0, z_1, \dots, z_n \in \mathbb{D}$$
 (distinct)

Given  $z_0, z_1, \ldots, z_n \in \mathbb{D}$  (distinct), find a Carathéodory function f such that

$$f(z_k) = c_k, \quad k = 0, 1, \dots, n$$



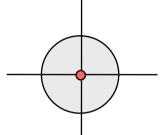


$$\operatorname{Re}\{f(z)\} \geq 0 \text{ in } \mathbb{D}$$

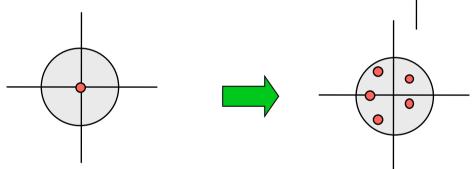
$$\int_{-\pi}^{\pi} \alpha_k(t) \operatorname{Re}\{f(e^{it})\} dt = c_k, \quad k = 0, 1, \dots, n$$

# Tuning by moving interpolation points

Covariance extension corresponds to N-P interpolation with all interpolation points at the origin (z = 0)

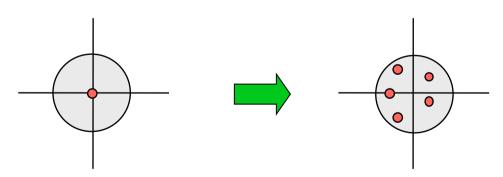


We moved the spectral zeros from the origin closer to the unit circle



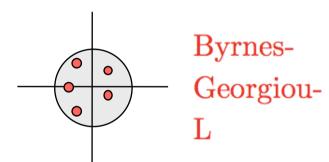
maximum entropy (LPC)

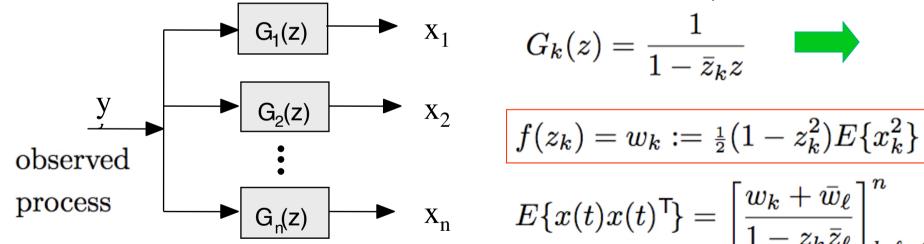
Next we tune by moving the interpolation points closer to the unit circle



# A tunable high resolution spectral estimator (THREE)

Zoom into a selected spectral band by moving interpolation points from the origin closer to the unit circle.





$$\mathbf{x}_{\mathrm{n}} \qquad E\{x(t)x(t)^{\mathsf{T}}\} = \left[rac{w_k + ar{w}_\ell}{1 - z_kar{z}_\ell}
ight]_{k,\ell=0}^n$$

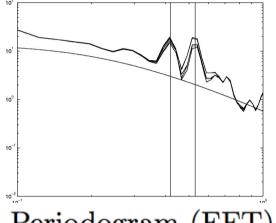
Two sets of tuning parameters: • filter bank poles

• spectral zeros (P)

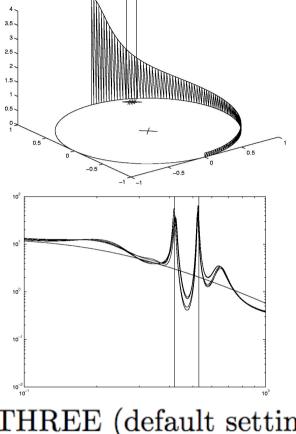
# Estimation of spectral lines in colored noise

separation between spectral lines = 0.11

five runs superimposed

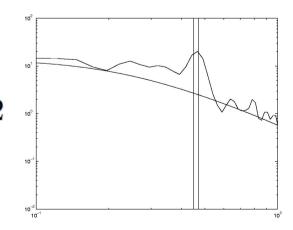


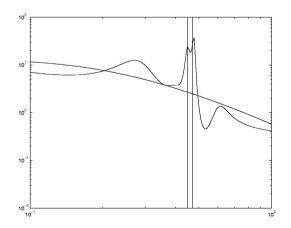
Periodogram (FFT)



THREE (default setting)

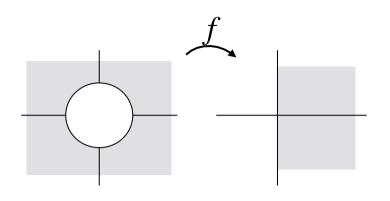
separation between spectral lines = 0.02

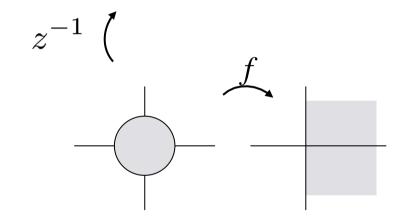




### Robust control







$$\frac{f}{1+}$$

### • positive real function

- f analytic for  $|z| \ge 1$
- $\operatorname{Re}\{f(z)\} > 0 \text{ for } |z| \ge 1$

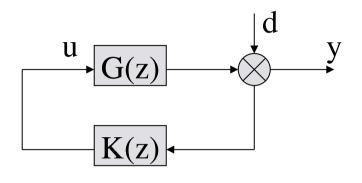
### Carathéodory function

- f analytic for  $|z| \le 1$
- $\operatorname{Re}\{f(z)\} > 0 \text{ for } |z| \le 1$

#### • Schur function

- f analytic for  $|z| \le 1$
- |f(z)| < 1 for  $|z| \le 1$

# Loop shaping in robust control



$$d \longrightarrow S(z) \longrightarrow S = (1 - GK)^{-1}$$

Sensitivity function

• Internal stability requires

$$S$$
 analytic in  $\mathbb{D}^c := \{z \mid |z| > 1\}$   
 $S(z_k) = 0$  at all unstable poles of  $G$ 

$$C(x_k)$$
 that the state of  $C$  is  $\mathbb{D}^C$ 

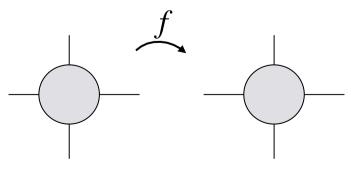
$$S(z_j) = 1$$
 at all zeros of  $G$  in  $\mathbb{D}^c$ 

• Disturbance attenuation requires

$$||S||_{\infty} \le \gamma$$

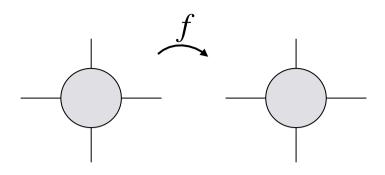
• We want  $\deg S$  to be small

There is a minimum bound  $\gamma_{\rm opt}$  but we choose  $\gamma > \gamma_{\rm opt}$  and define  $f(z) := \frac{1}{\gamma} S(z^{-1})$ 



Nevanlinna-Pick interpolation for Schur functions

$$f(z_k) = w_k, \quad k = 0, 1, \dots, n$$



class of Schur functions S

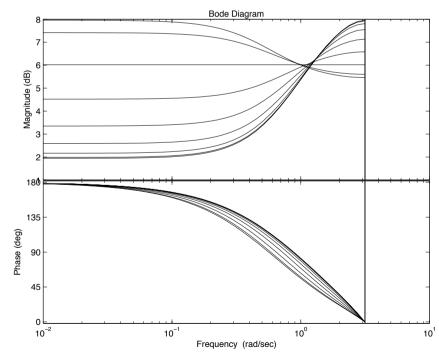
The interpolants of degree at most n are parameterized by the spectral zeros  $(\sigma)$  in a 1-1 fashion

Example 
$$G(z) = \frac{1}{z-2}$$
 
$$S(2) = 0, S(\infty) = 1$$

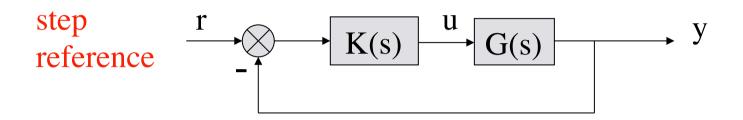
Find all S of degree at most n = 1.

$$S(z) = \frac{z-2}{z-a}, -1 < a < 1$$

$$\gamma = 2.5 > \gamma_{\rm opt} = \min \|S\|_{\infty} = 2$$



# Example: Sensitivity shaping



$$G(s) = \frac{-6.4750s^2 + 4.0302s + 175.7700}{s(5s^3 + 3.5682s^2 + 139.5021s + 0.0929)}$$

Doyle, Francis Tannenbaum

Design a strictly proper *K* so that the closed-loop system is

- internally stable and
- satisfies the specifications:
- settling time at most 8 seconds
- overshoot at most 10%
- $|u(t)| \le 0.5$

Try to achieve 
$$S_{ideal} = \frac{s(s+1.2)}{s^2+1.2s+1}$$
 (DFT)

Internal stability requires  $|\deg S \le 4|$ 

$$\deg S \le 4$$

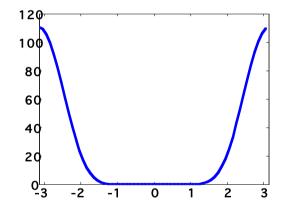
$$S(0) = S'(\infty) = 0$$
 (unstable plant poles)

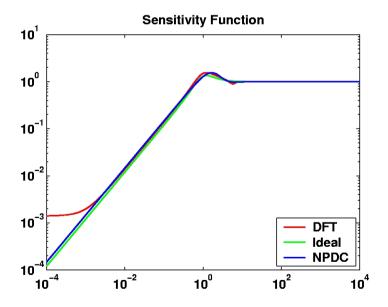
$$S(\infty) = S(5.5308) = 1$$
 (nonminimum-phase plant zeros)

$$S''(\infty) = 0$$
 (strictly proper controller)



$$|\sigma|^2$$

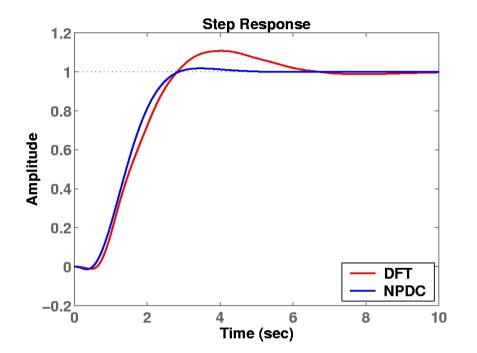


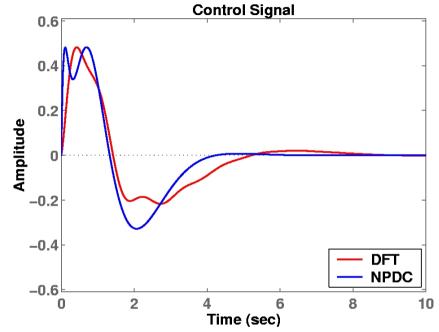


DFT: deg K = 8

NPDC: deg K = 4

A. Blomqvist and R. Nagamune





- settling time at most 8 seconds
- overshoot at most 10%
- $|u(t)| \le 0.5$

DFT: deg C = 8NPDC: deg C = 4

## Multidimensional moment problems

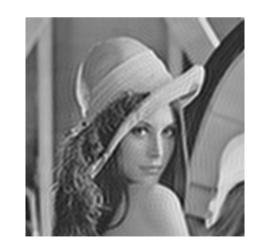
$$\int_{K} \alpha_k d\mu = c_k, \quad k = 1, 2, \dots, n$$

- $d\mu$  nonnegative measure on a compact subset K of  $\mathbb{R}^d$
- $\alpha_1, \alpha_2, \dots \alpha_n$  linearly independent basis functions defined on K

#### Image compression







### Model reduction

### Original system:

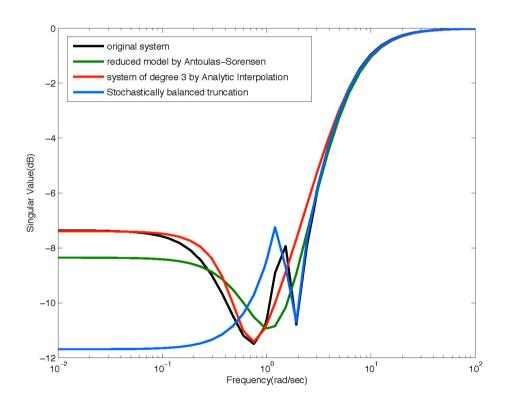
$$G(s) = \frac{s^5 + 3s^4 + 6s^3 + 9s^2 + 7s + 3}{s^5 + 7s^4 + 14s^3 + 21s^2 + 23s + 7}$$

#### Antoulas-Sorensen:

$$\hat{G}(s) = \frac{s^3 + 2.553s^2 + 2.906s + 1.173}{s^3 + 6.681s^2 + 8.459s + 3.07}$$

### Global-analysis approach:

$$f(s) = \frac{1.002s^3 + 2.84s^2 + 1.927s + 0.8978}{s^3 + 7.298s^2 + 6.084s + 2.099}$$

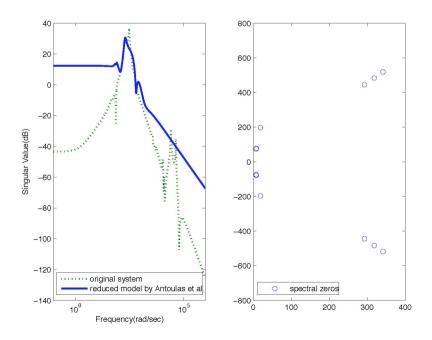


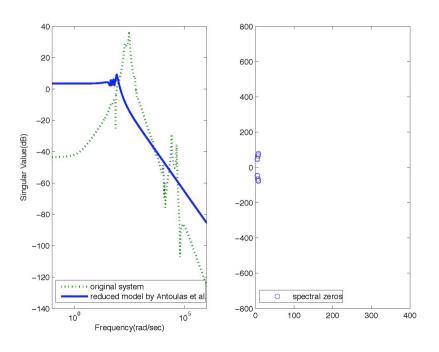
# A large-scale problem: A CD player

Model reduction:

$$\deg G = 120 \longrightarrow \deg \hat{G} = 12$$

#### Antoulas-Sorensen solutions:



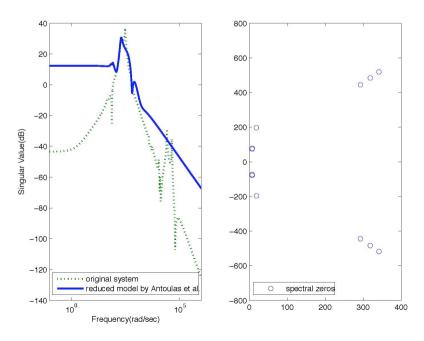


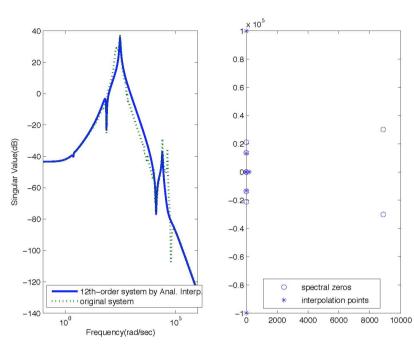
# Global-analysis solution

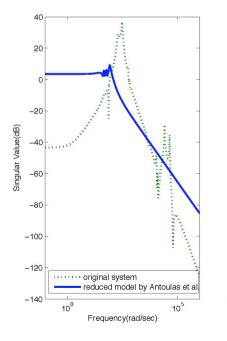


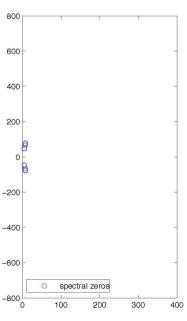
# Antoulas-Sorensen solutions











### Conclusions

An enhanced theory for generalized moment problems that incorporates rationality constraints prescribed by applications.

- Complete parameterizations of solutions with smooth tuning strategies.
- A global analysis approach that studies the class of solutions as a whole.
- Convex optimization for determining solutions.